

Plug N Drive™ Integrated Power  
 Module for Appliance Motor Drive

## Description

International Rectifier's IRAMS10UP60A is an Integrated Power Module developed and optimized for electronic motor control in appliance applications such as washing machines and refrigerators. Plug N Drive technology offers an extremely compact, high performance AC motor-driver in a single isolated package for a very simple design.

A built-in temperature monitor and over-temperature/over-current protection, along with the short-circuit rated IGBTs and integrated under-voltage lockout function, deliver high level of protection and fail-safe operation.

The integration of the bootstrap diodes for the high-side driver section, and the single polarity power supply required to drive the internal circuitry, simplify the utilization of the module and deliver further cost reduction advantages.

## Features

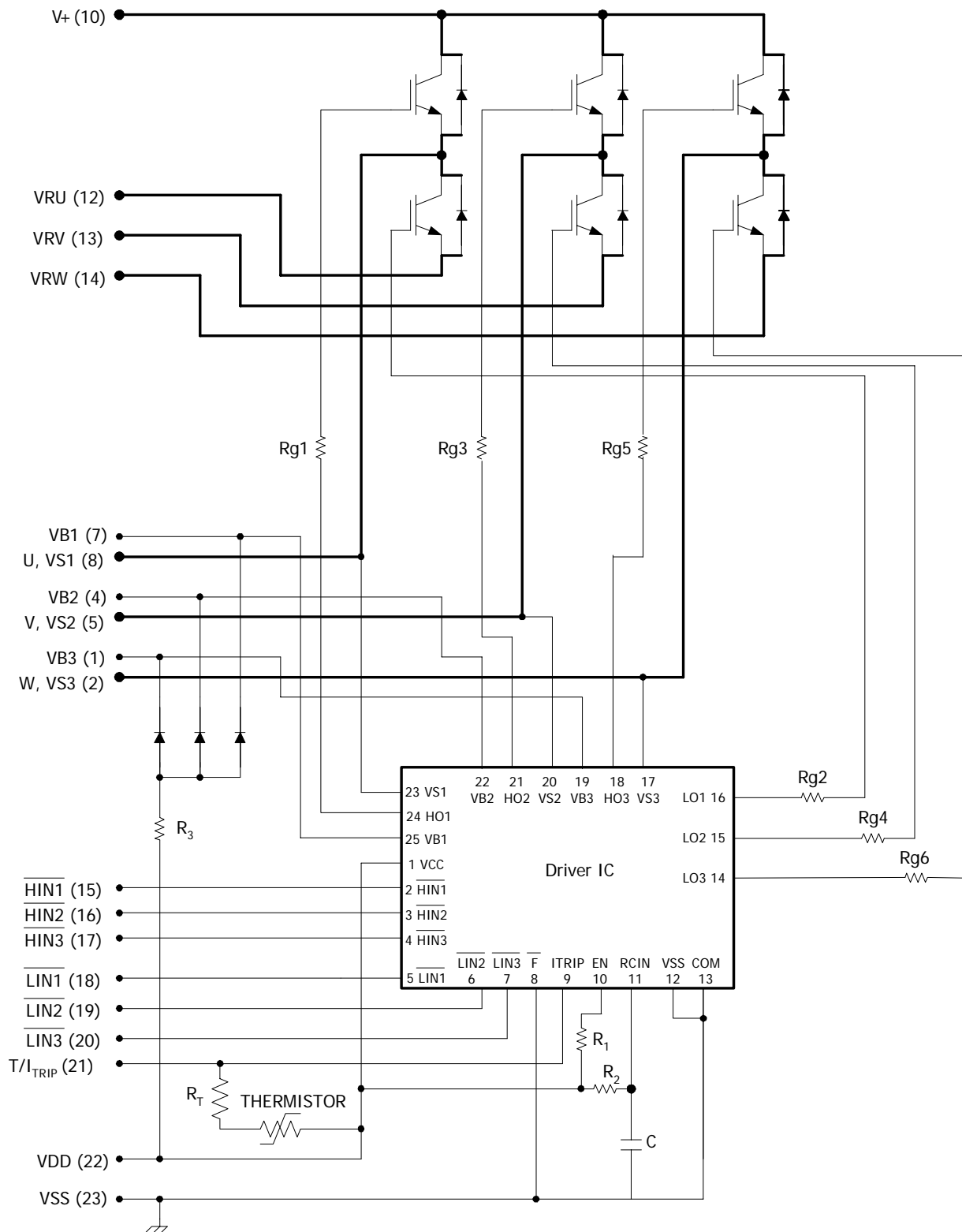
- Integrated Gate Drivers and Bootstrap Diodes.
- Temperature Monitor
- Temperature and Overcurrent shutdown
- Fully Isolated Package.
- Low VCE (on) Non Punch Through IGBT Technology
- Under-voltage lockout for all channels
- Matched propagation delay for all channels
- Low side IGBT emitter pins for current control
- Schmitt-triggered input logic
- Cross-conduction prevention logic
- Lower di/dt gate driver for better noise immunity



## Absolute Maximum Ratings

Parameter	Description	Max. Value	Units
$V_{CES}$	Maximum IGBT Blocking Voltage	600	V
$V^+$	Positive Bus Input Voltage	450	
$I_O @ T_C = 25^\circ C$	RMS Phase Current	10	A
$I_O @ T_C = 100^\circ C$	RMS Phase Current	5	
$I_{pk}$	Maximum Peak Phase Current (tp<100ms)	15	
$F_p$	Maximum PWM Carrier Frequency	20	kHz
$P_d$	Maximum Power dissipation per Phase	20	W
$V_{iso}$	Isolation Voltage (1min)	2000	$V_{RMS}$
$T_J$ (IGBT & Diodes)	Operating Junction temperature Range	-40 to +150	°C
$T_J$ (Driver IC)	Operating Junction temperature Range	-40 to +150	
T	Mounting torque Range (M3 screw)	0.8 to 1.0	Nm

## Internal Electrical Schematic - IRAMS10UP60A



## Inverter Section Electrical Characteristics @ $T_J = 25^\circ\text{C}$

Symbol	Parameter	Min	Typ	Max	Units	Conditions
$V_{(BR)CES}$	Collector-to-Emitter Breakdown Voltage	600	---	---	V	$V_{IN}=0V, I_C=20\mu A$
$\Delta V_{(BR)CES} / \Delta T$	Temperature Coeff. Of Breakdown Voltage	---	0.57	---	V/ $^\circ\text{C}$	$V_{IN}=0V, I_C=1.0mA$ ( $25^\circ\text{C} - 150^\circ\text{C}$ )
$V_{CE(ON)}$	Collector-to-Emitter Saturation Voltage	---	1.7	2.0	V	$I_C=5A, T_J=25^\circ\text{C}, V_{DD}=15V$
		---	2.0	2.4		$I_C=5A, T_J=150^\circ\text{C}$
$I_{CES}$	Zero Gate Voltage Collector Current-to-Emitter	---	5	15	$\mu A$	$V_{IN}=5V, V^+=600V$
		---	10	40		$V_{IN}=5V, V^+=600V, T_J=150^\circ\text{C}$
$I_{Ik\_module}$	Zero Gate Phase-to-Phase Current	--	--	50	$\mu A$	$V_{IN}=5V, V^+=600V$
$V_{FM}$	Diode Forward Voltage Drop	---	1.8	2.35	V	$I_C=5A$
		---	1.3	1.7		$I_C=5A, T_J=150^\circ\text{C}$

## Inverter Section Switching Characteristics

Symbol	Parameter	Min	Typ	Max	Units	Conditions
$E_{on}$	Turn-On Switching Loss	---	200	235	$\mu J$	$I_C=5A, V^+=400V$
$E_{off}$	Turn-Off Switching Loss	---	75	100		$V_{DD}=15V, L=1mH$
$E_{tot}$	Total Switching Loss	---	275	335		See CT1 $T_J=25^\circ\text{C}$
$E_{on}$	Turn-on Switching Loss	---	300	360	$\mu J$	$T_J=150^\circ\text{C}$
$E_{off}$	Turn-off Switching Loss	---	135	165		Energy losses include "tail" and diode reverse recovery
$E_{tot}$	Total Switching Loss	---	435	525		
$E_{rec}$	Diode Reverse Recovery energy	---	30	40	$\mu J$	$T_J=150^\circ\text{C}, V^+=400V, V_{DD}=15V, I_F=5A, L=1mH$
$t_{rr}$	Diode Reverse Recovery time	---	100	145	ns	
RBSOA	Reverse Bias Safe Operating Area	FULL SQUARE				$T_J=150^\circ\text{C}, I_C=5A, V_p=600V$ $V^+=480V, V_{DD}=+15V$ to 0V See CT3
SCSOA	Short Circuit Safe Operating Area	10	---	---	$\mu s$	$T_J=150^\circ\text{C}, V_p=600V, V^+=360V, V_{DD}=+15V$ to 0V See CT2

## Thermal Resistance

Symbol	Parameter	Min	Typ	Max	Units	Conditions
$R_{th(J-C)}$	Junction to case thermal resistance, each IGBT under inverter operation.	---	4.2	4.7	$^\circ\text{C}/W$	Flat, greased surface. Heatsink compound thermal conductivity - 1W/mK
$R_{th(J-C)}$	Junction to case thermal resistance, each Diode under inverter operation.	---	5.5	6.5	$^\circ\text{C}/W$	
$R_{th(C-S)}$	Thermal Resistance case to sink	---	0.1	---	$^\circ\text{C}/W$	

## Absolute Maximum Ratings Driver Function

Absolute Maximum Ratings indicate substaines limits beyond which damage to the device may occur. All voltage parameters are absolute voltages referenced to  $V_{SS}$  (Note 1)

Symbol	Definition	Min	Max	Units
$V_{S1,2,3}$	High Side offset voltage	-0.3	600	V
$V_{B1,2,3}$	High Side floating supply voltage	-0.3	20	V
$V_{DD}$	Low Side and logic fixed supply voltage	-0.3	20	V
$V_{IN}$	Input voltage LIN, HIN, T/I <sub>TRIP</sub>	-0.3	7	V
$T_J$	Juction Temperature	-40	150	°C

## Recommended Operating Conditions Driver Function

The Input/Output logic timing diagram is shown in Figure 1. For proper operation the device should be used within the recommended conditions. All voltages are absolute referenced to  $V_{SS}$ . The  $V_S$  offset is tested with all supplies biased at 15V differential (Note 1). All input pin ( $V_{IN}$ ) and  $I_{TRIP}$  are clamped with a 5.2V zener diode and pull-up resistor to  $V_{DD}$

Symbol	Definition	Min	Max	Units
$V_{B1,2,3}$	High side floating supply voltage	$V_S+12$	$V_S+20$	V
$V_{S1,2,3}$	High side floating supply offset voltage	Note 2	450	
$V_{DD}$	Low side and logic fixed supply voltage	12	20	V
$V_{ITRIP}$	T/I <sub>TRIP</sub> input voltage	$V_{SS}$	$V_{SS}+5$	
$V_{IN}$	Logic input voltage LIN, HIN	$V_{SS}$	$V_{SS}+5$	V

## Static Electrical Characteristics Driver Function

$V_{BIAS}$  ( $V_{CC}$ ,  $V_{BS1,2,3}$ )=15V, unless otherwise specified. The  $V_{IN}$  and  $I_{IN}$  parameters are referenced to  $V_{SS}$  and are applicable to all six channels. (Note 1)

Symbol	Definition	Min	Typ	Max	Units
$V_{IN,th+}$	Positive going input threshold	3.0	---	---	V
$V_{IN,th-}$	Negative going input threshold	---	---	0.8	V
$V_{CCUV+}$ $V_{BSUV+}$	$V_{CC}$ and $V_{BS}$ supply undervoltage Positive going threshold	10.6	11.1	11.6	V
$V_{CCUV-}$ $V_{BSUV-}$	$V_{CC}$ and $V_{BS}$ supply undervoltage Negative going threshold	10.4	10.9	11.4	V
$V_{CCUVH}$ $V_{BSUVH}$	$V_{CC}$ and $V_{BS}$ supply undervoltage $I_{lockout}$ hysteresis	---	0.2	---	V
$I_{OBS}$	Quiescent $V_{BS}$ supply current	---	70	120	μA
$I_{OCC}$	Quiscent $V_{CC}$ supply current	---	1.6	2.3	mA
$I_{LK}$	Offset Supply Leakage Current	---	---	50	μA
$I_{IN+}$	Input bias current (OUT=LO)	---	100	220	μA
$I_{IN-}$	Input bias current (OUT=HI)	---	200	300	μA
$V(I_{TRIP})$	$I_{TRIP}$ threshold Voltage (OUT=HI or OUT=LO)	3.85	4.3	4.75	V

## Dynamic Electrical Characteristics

$V_{DD}=V_{BS}=V_{BIAS}=15V$ ,  $I_o=1A$ ,  $V_D=9V$ ,  $PWM_{IN}=2kHz$ ,  $V_{IN\_ON}=V_{IN\_th+}$ ,  $V_{IN\_OFF}=V_{IN\_th-}$

$T_A=25^{\circ}C$ , unless otherwise specified

Symbol	Definition	Min	Typ	Max	Units
$T_{ON}$	Input to output propagation turn-on delay time (see fig. 11)	-	470	-	ns
$T_{OFF}$	Input to output propagation turn-off delay time (see fig. 11)	-	615	-	ns
$D_T$	Dead Time	-	300	-	ns
$I/T_{Trip}$	$T/I_{Trip}$ to six switch to turn-off propagation delay (see fig. 2)	-	750	-	ns
$T_{FCLTRL}$	Post $I_{Trip}$ to six switch to turn-off clear time (see fig. 2)	-	9	-	ms

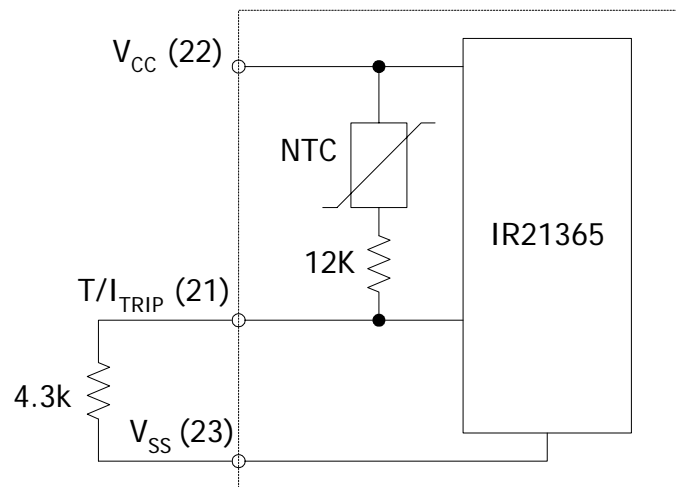
## Internal NTC - Thermistor Characteristics

Parameter	Typ	Units	Conditions
$R_{25}$ Resistance	100 +/- 5%	k $\Omega$	$T_C = 25^{\circ}C$
$R_{125}$ Resistance	2.522 + 17.3 % /- 14.9%	k $\Omega$	$T_C = 125^{\circ}C$
B B-Constant (25-50 $^{\circ}C$ )	4250 +/- 3%	k	$R_2 = R_1 e^{[B(1/T_2 - 1/T_1)]}$
Temperature Range	-40 / 125	$^{\circ}C$	
Typ. Dissipation constant	1	mW/ $^{\circ}C$	$T_C = 25^{\circ}C$

Note 1: For more details, see IR21365 data sheet

Note 2: Logic operational for  $V_S$  from  $V-5V$  to  $V+600V$ . Logic stata held for  $V_S$  from  $V-5V$  to  $V-V_{BS}$ . (Please refer to DT97-3 for more details)

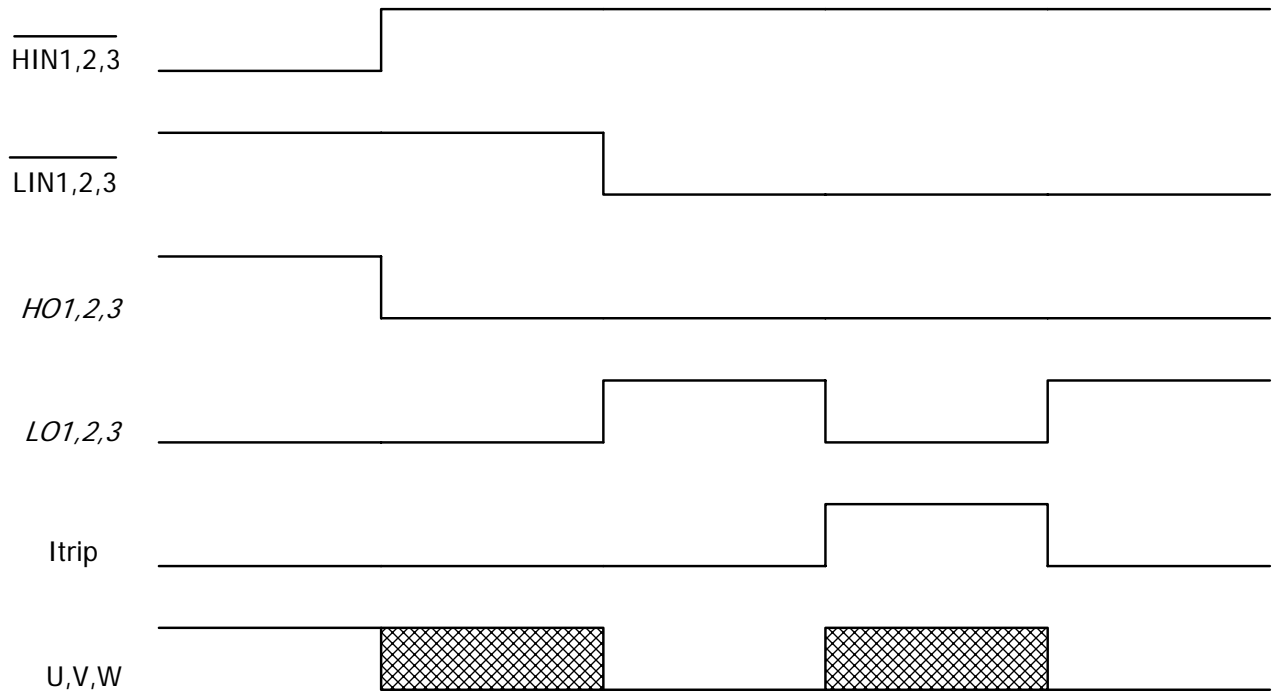
## Thermistor Built-in IRAMS10UP60A



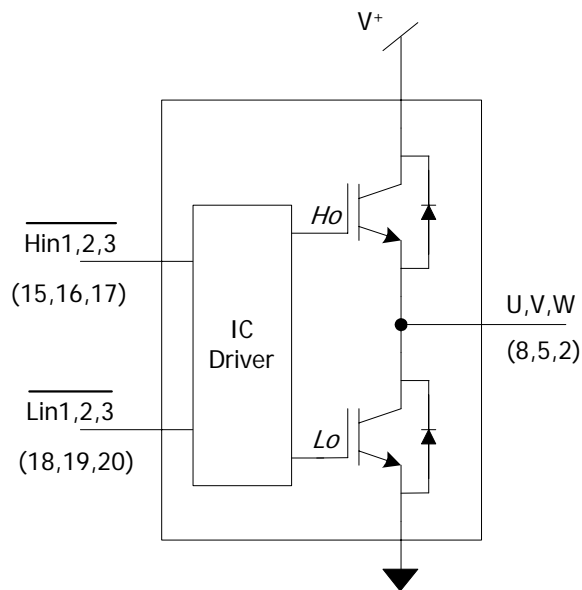
Note 3: The Maximum recommended sense voltage at the  $T/I_{TRIP}$  terminal under normal operating conditions is 3.3V.

# IRAMS10UP60A

Figure1. Input/Output Timing Diagram

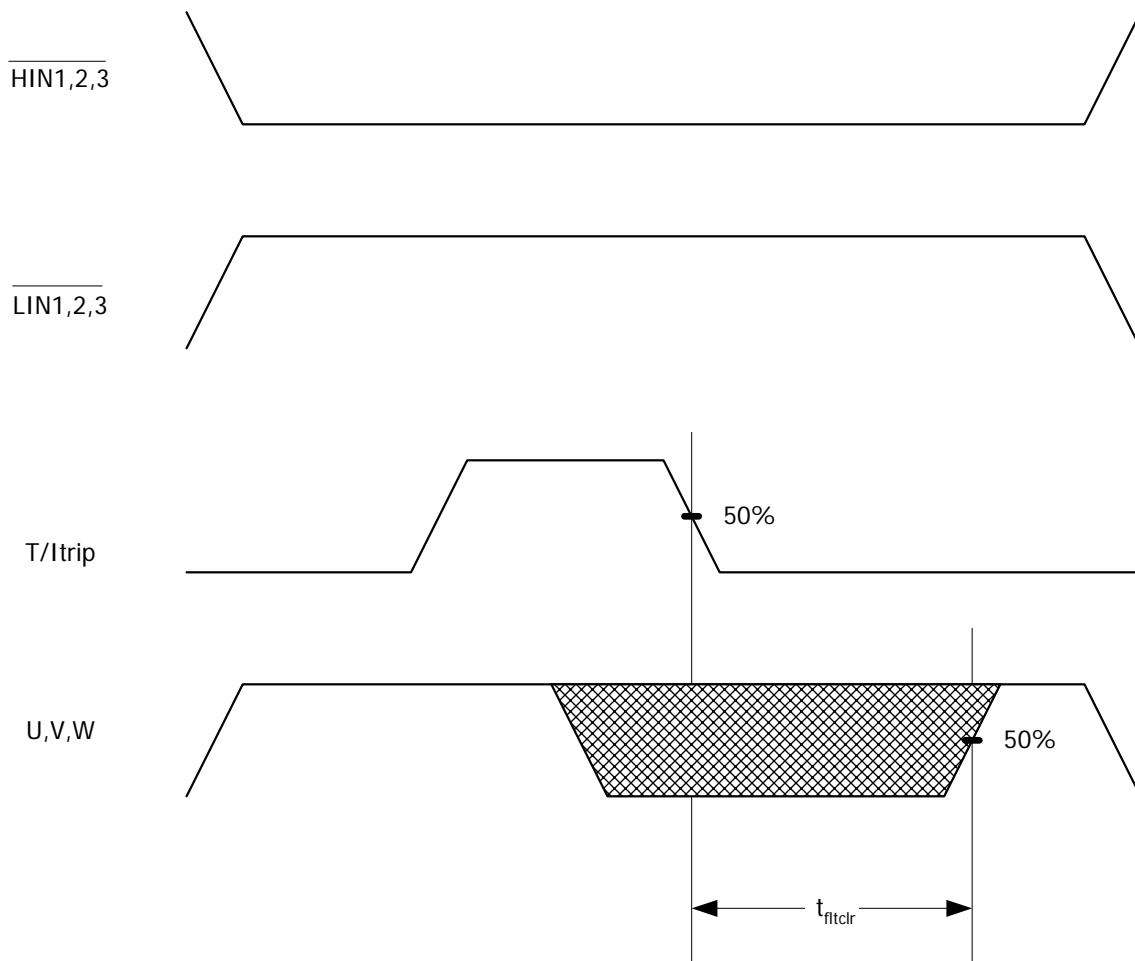


Note 4: The shaded area indicates that both high-side and low-side switches are off and therefore the half-bridge output voltage would be determined by the direction of current flow in the load.



$I_{trip}$	$\overline{HIN1,2,3}$	$\overline{LIN1,2,3}$	$U,V,W$
0	0	1	$V_{bus}$
0	1	0	0
0	1	1	X
1	X	X	X

Figure 2.  $T/I_{Trip}$  Timing Waveform



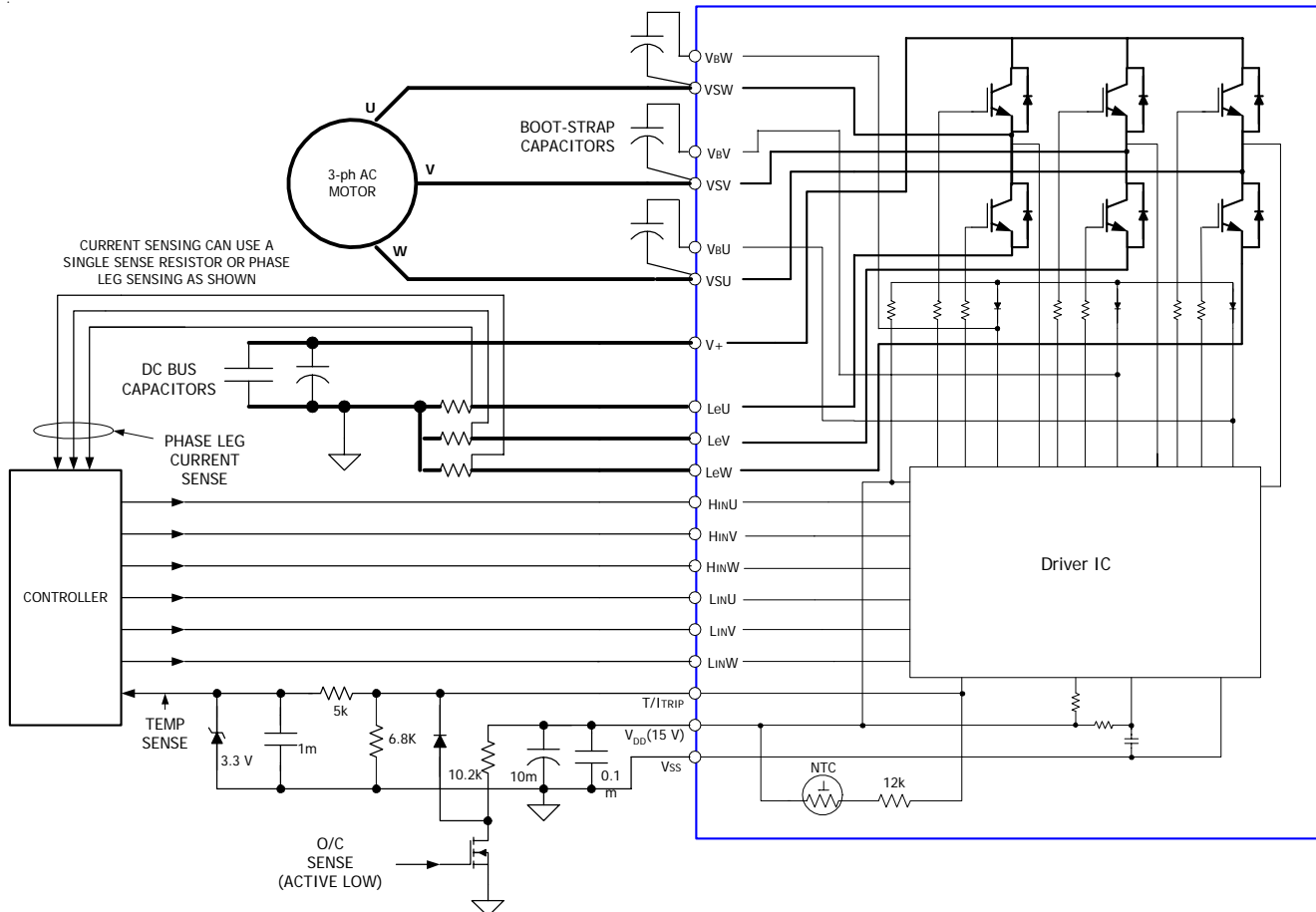
Note 5: The shaded area indicates that both high-side and low-side switches are off and therefore the half-bridge output voltage would be determined by the direction of current flow in the load.

## Module Pin-Out Description

Pin	Name	Description
1	VB3	High Side Floating Supply Voltage 3
2	W, VS3	Output 3 - High Side Floating Supply Offset Voltage
3	na	none
4	VB2	High Side Floating Supply voltage 2
5	V, VS2	Output 2 - High Side Floating Supply Offset Voltage
6	na	none
7	VB1	High Side Floating Supply voltage 1
8	U, VS1	Output 1 - High Side Floating Supply Offset Voltage
9	na	none
10	V+	Positive Bus Input Voltage
11	na	none
12	LE1	Low Side Emitter Connection - Phase 1
13	LE2	Low Side Emitter Connection - Phase 2
14	LE3	Low Side Emitter Connection - Phase 3
15	HIN1	Logic Input High Side Gate Driver - Phase 1
16	HIN2	Logic Input High Side Gate Driver - Phase 2
17	HIN3	Logic Input High Side Gate Driver - Phase 3
18	LIN1	Logic Input Low Side Gate Driver - Phase 1
19	LIN2	Logic Input Low Side Gate Driver - Phase 2
20	LIN3	Logic Input Low Side Gate Driver - Phase 3
21	T/Itrip	Temperature Monitor and Shut-down Pin
22	VCC	+15V Main Supply
23	VSS	Negative Main Supply



## Typical Application Connection IRAMS10UP60A



1. Electrolytic bus capacitors should be mounted as close to the module bus terminals as possible to reduce ringing and EMI problems. Additional high frequency ceramic capacitor mounted close to the module pins will further improve performance.
2. In order to provide good decoupling between  $V_{CC}$ -Gnd and  $V_B$ - $V_{SS}$  terminals, the capacitors shown connected between these terminals should be located very close to the module pins. Additional high frequency capacitors, typically  $0.1\mu\text{F}$ , are strongly recommended.
3. Value of the boot-strap capacitors depends upon the switching frequency. Their selection should be made based on IR design tip DN 98-2a, application note AN-1044 or Figure 9.
4. Low inductance shunt resistors should be used for phase leg current sensing. Similarly, the length of the traces between pins 12, 13 and 14 to the corresponding shunt resistors should be kept as small as possible.
5. Over-current sense signal can be obtained from external hardware detecting excessive instantaneous current in inverter.

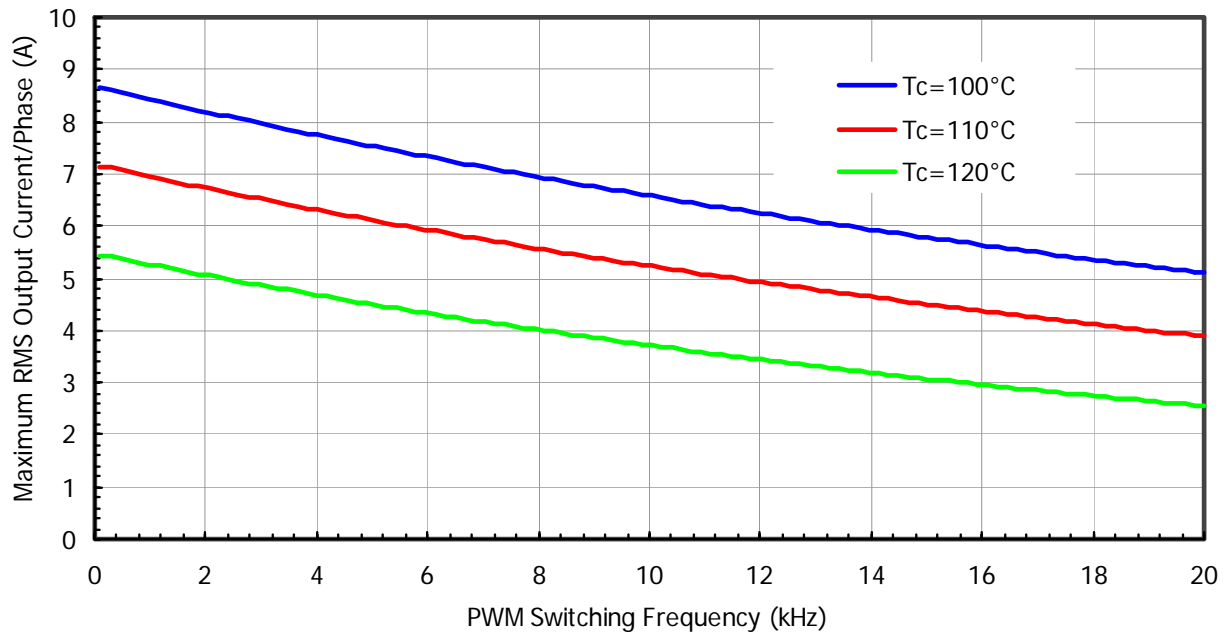


Figure 3. Maximum sinusoidal phase current as function of switching frequency  
 $V_{BUS}=400V$ ,  $T_J=150^\circ C$ , Modulation Depth=0.8, PF=0.6

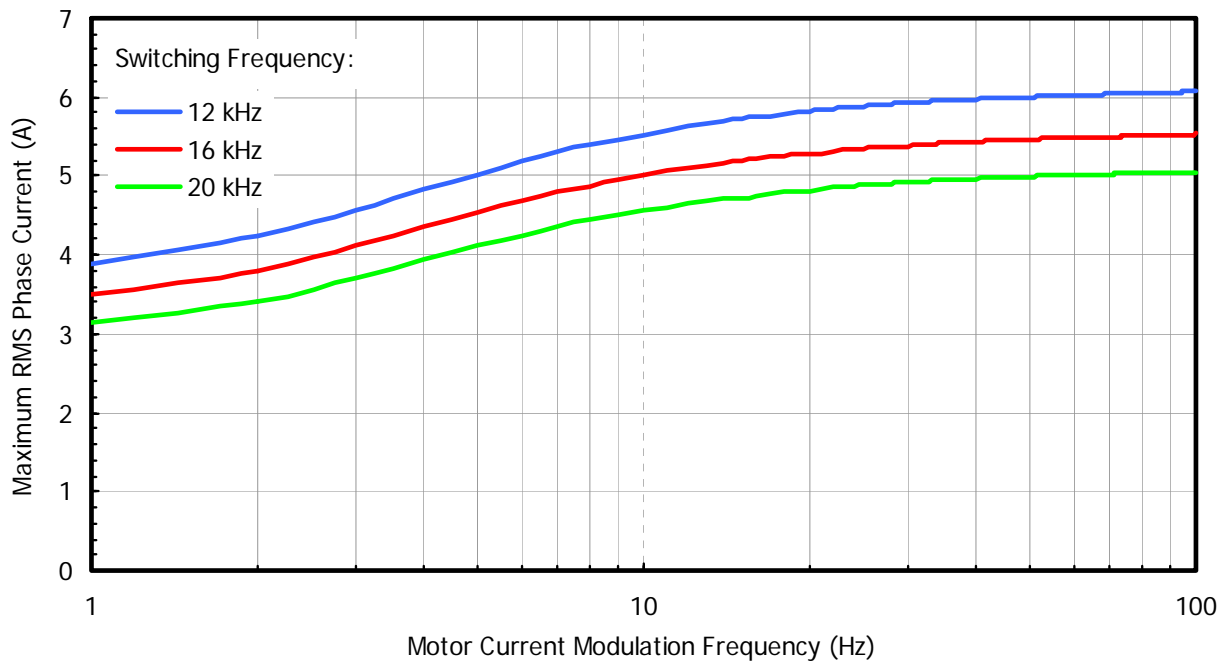


Figure 4. Maximum sinusoidal phase current as function of modulation frequency  
 $V_{BUS}=400V$ ,  $T_J=150^\circ C$ ,  $T_c=100^\circ C$ , Modulation Depth=0.8, PF=0.6

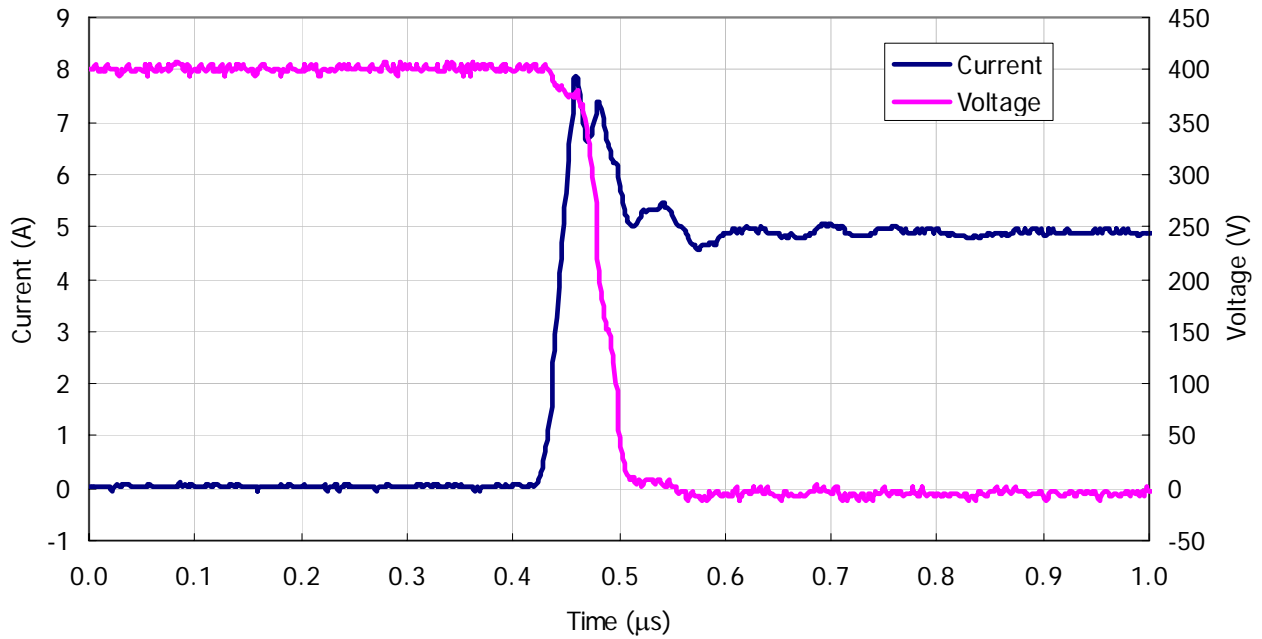


Figure 5. IGBT Turn-on. Typical turn-on waveform @ $T_j=125^\circ\text{C}$ ,  $V_{\text{BUS}}=400\text{V}$

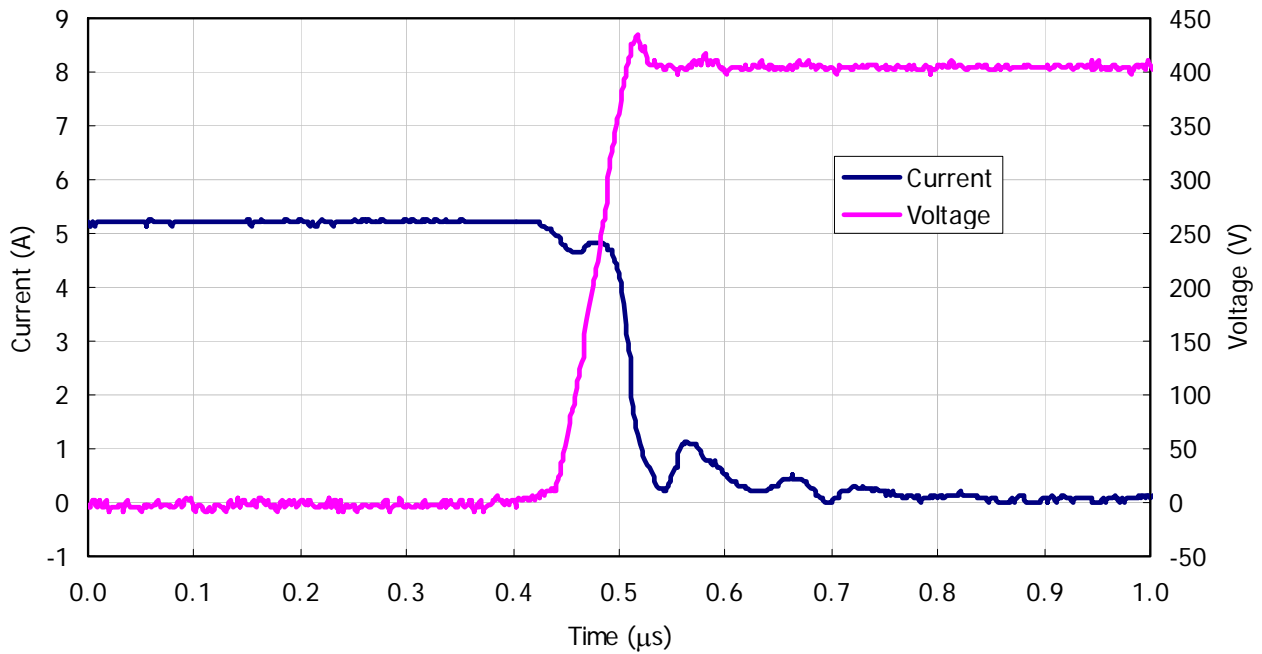


Figure 6. IGBT Turn-off. Typical turn-off waveform @ $T_j=125^\circ\text{C}$ ,  $V_{\text{BUS}}=400\text{V}$

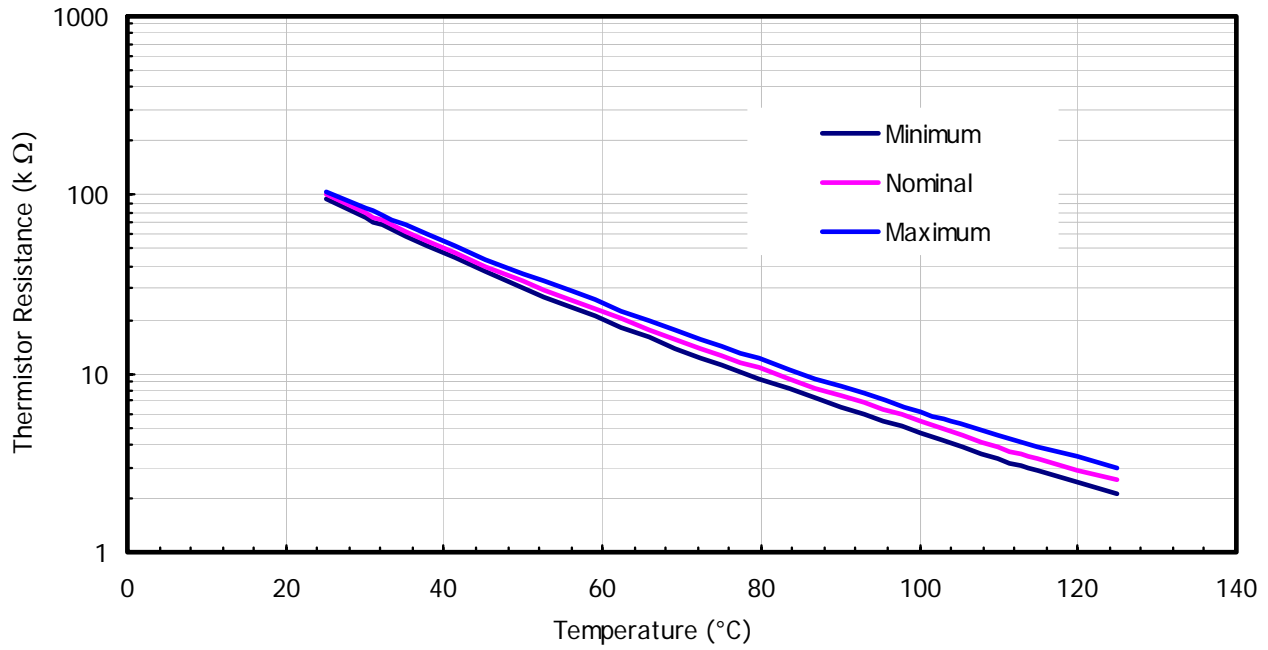


Figure 7. Variation of thermistor resistance with temperature

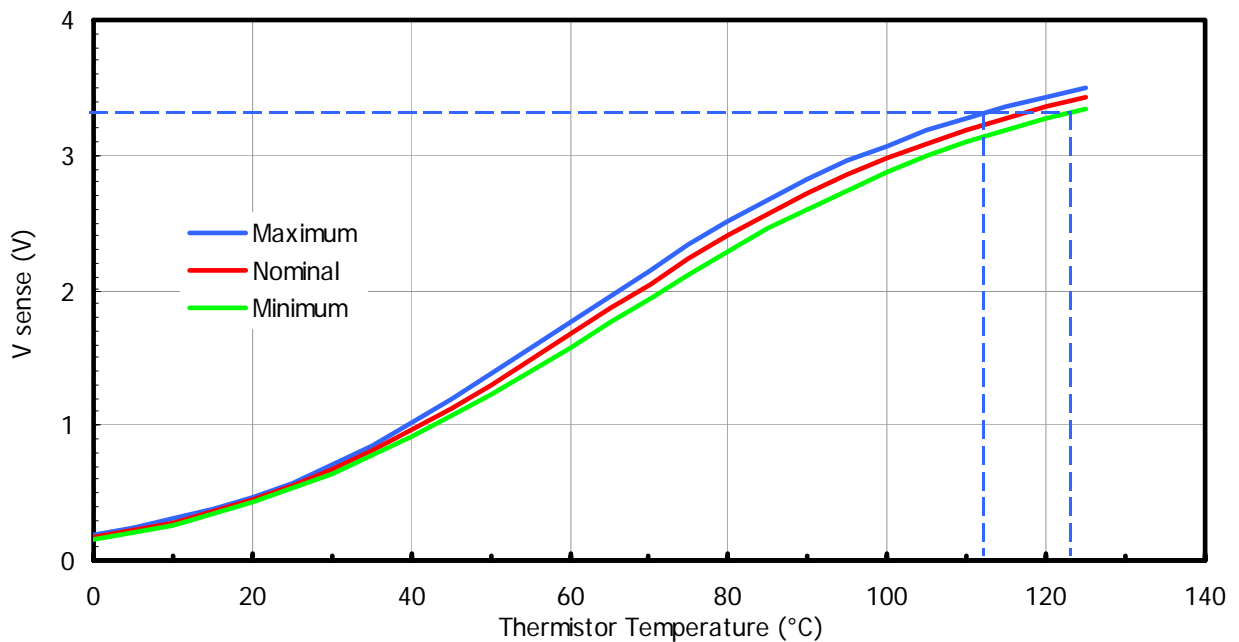


Figure 8. Variation of temperature sense voltage with thermistor temperature using external bias resistance of  $4.3\text{k}\Omega$ ,  $V_{CC}=15\text{V}$

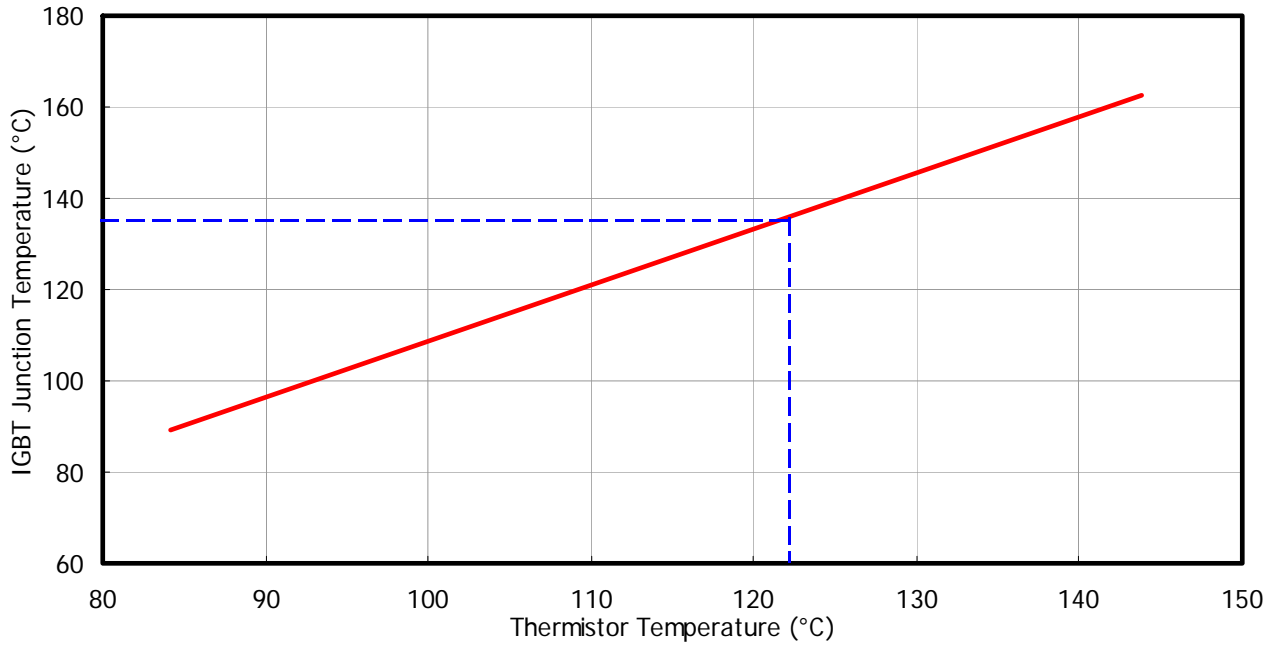


Figure 9. Estimated maximum IGBT junction temperature with thermistor temperature

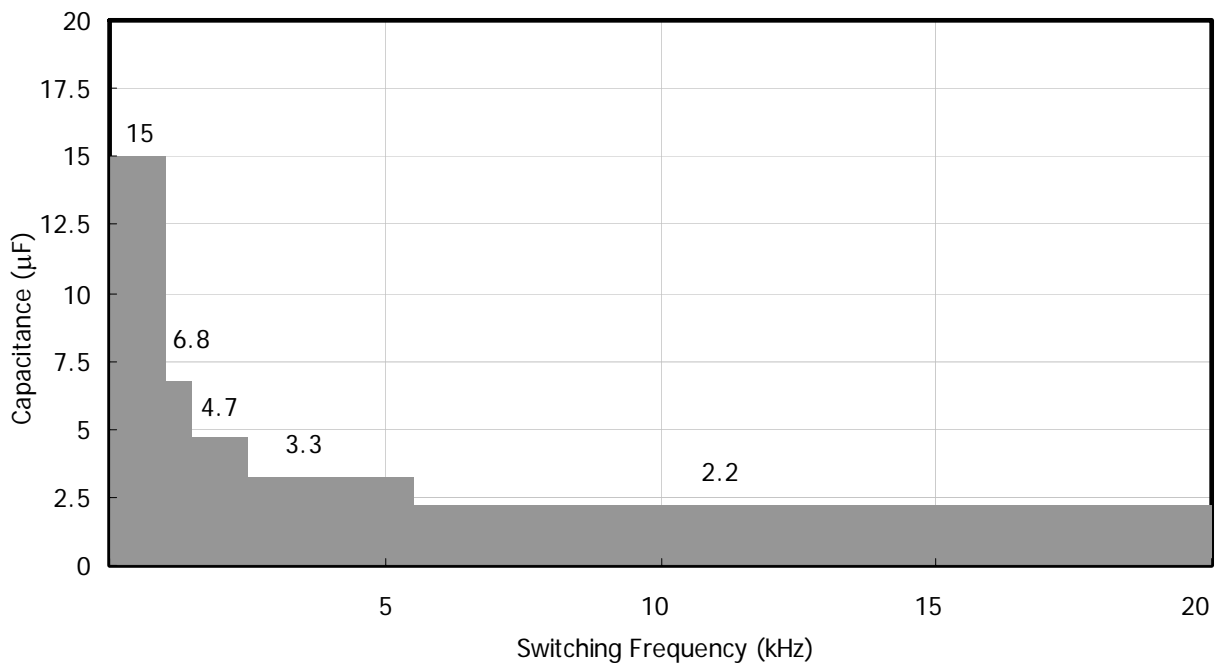


Figure 10. Recommended minimum Bootstrap Capacitor value Vs Switching Frequency

Figure 11. Switching Parameter Definitions

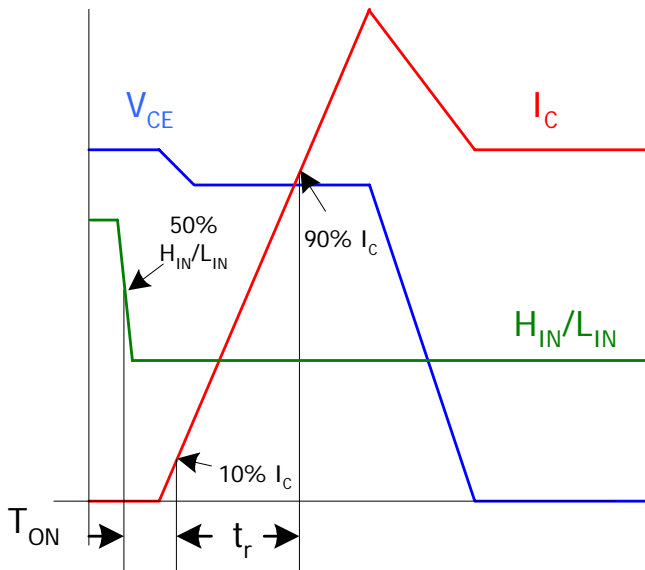


Figure 11a. Input to Output propagation turn-on delay time

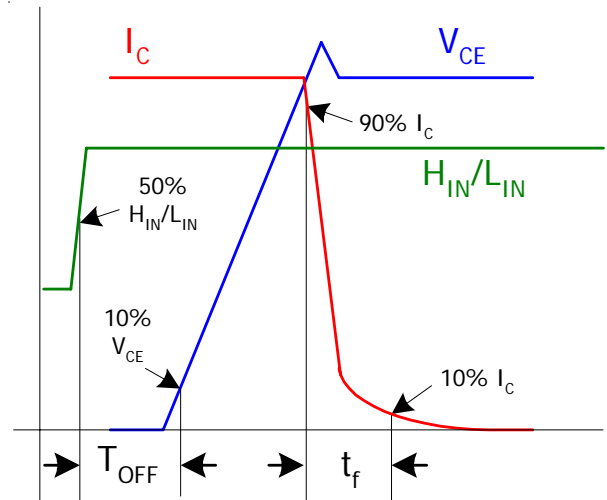


Figure 11b. Input to Output propagation turn-off delay time

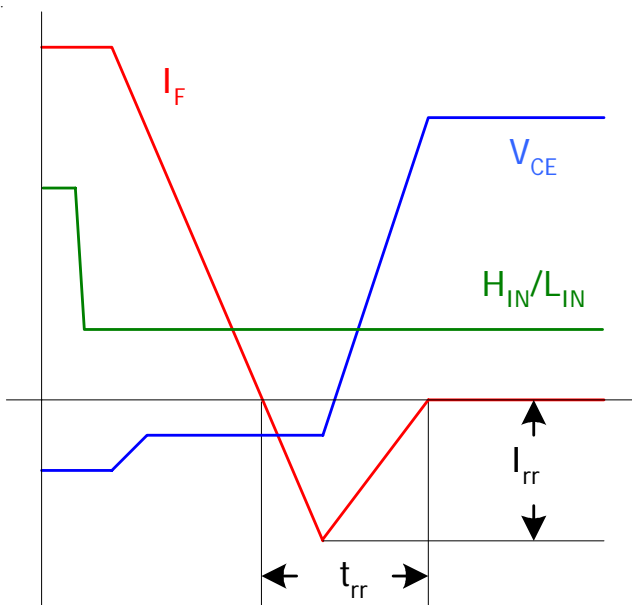


Figure 11c. Diode Reverse Recovery

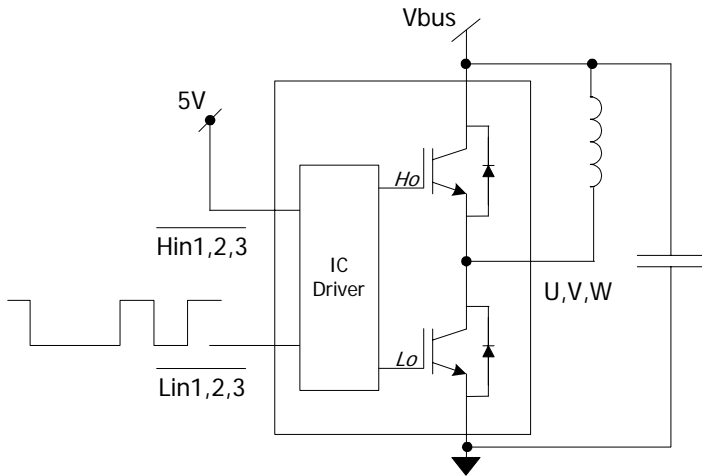


Figure CT1. Switching Loss Circuit

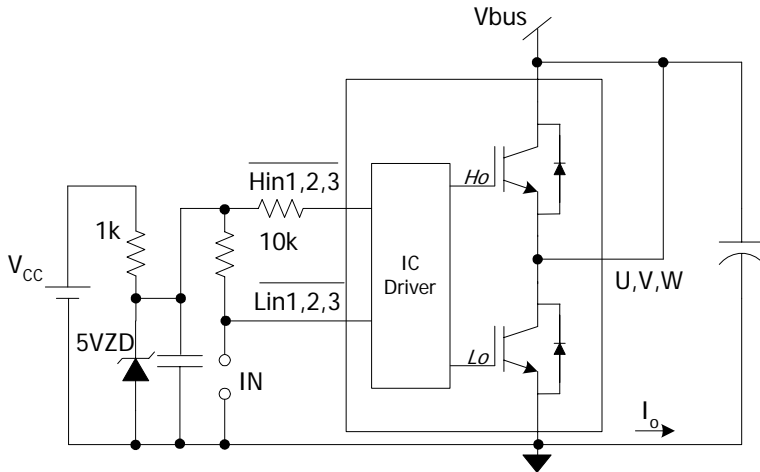
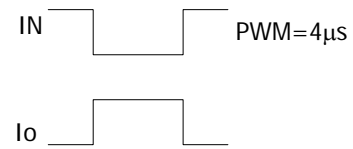


Figure CT2. S.C.SOA Circuit



$V_p$  = Peak Voltage on the IGBT die

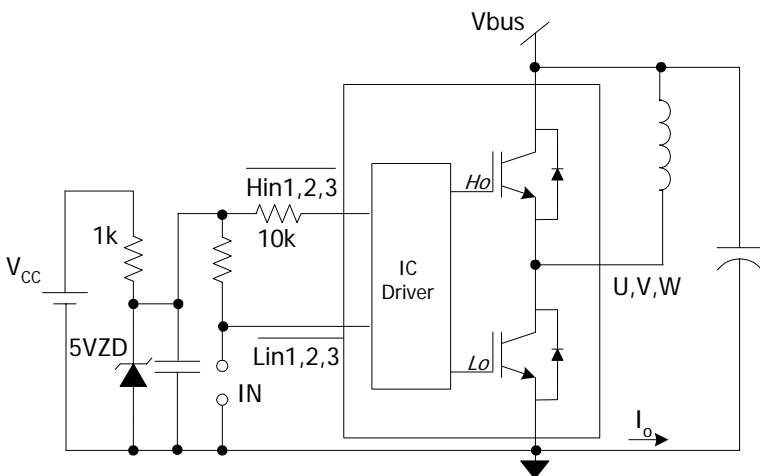
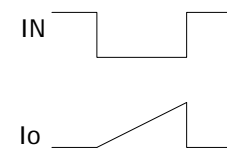
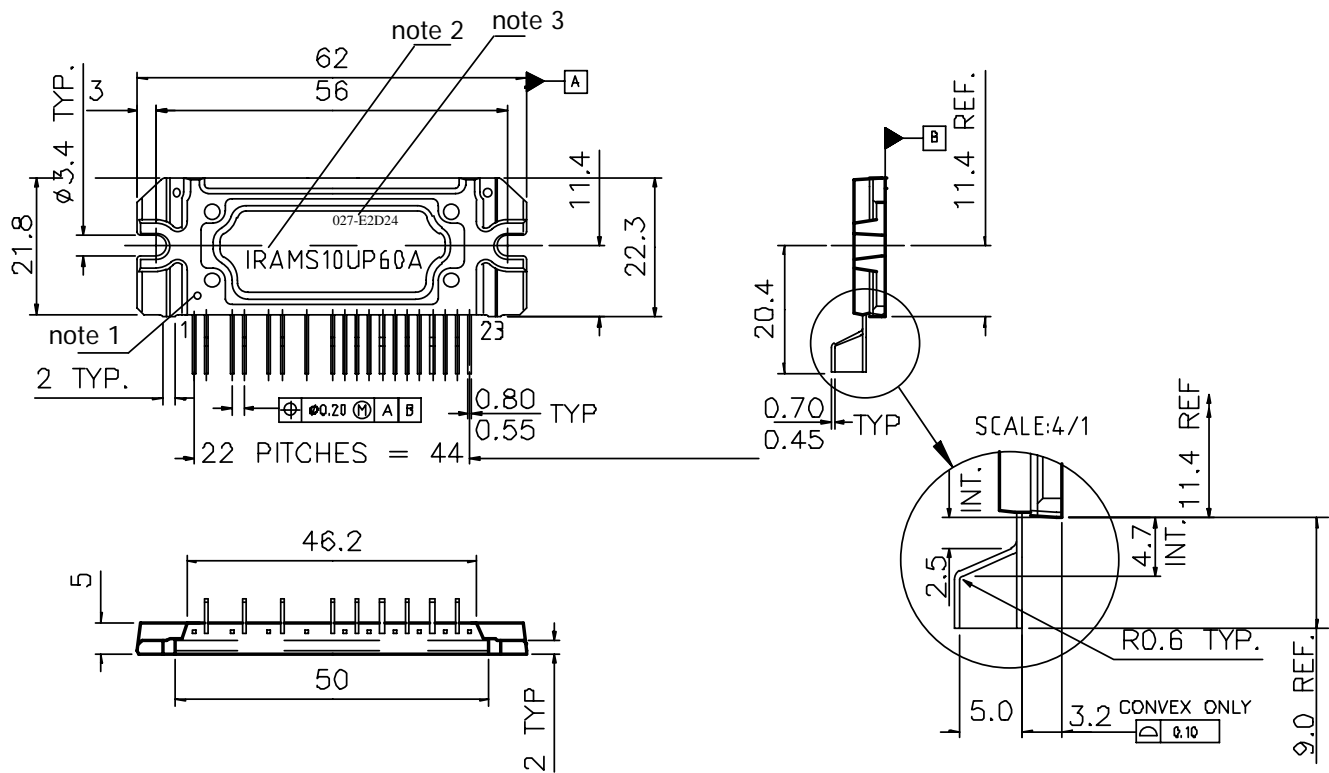


Figure CT3. R.B.SOA Circuit



$V_p$  = Peak Voltage on the IGBT die

## Package Outline



## Standard pin leadforming option

Notes:

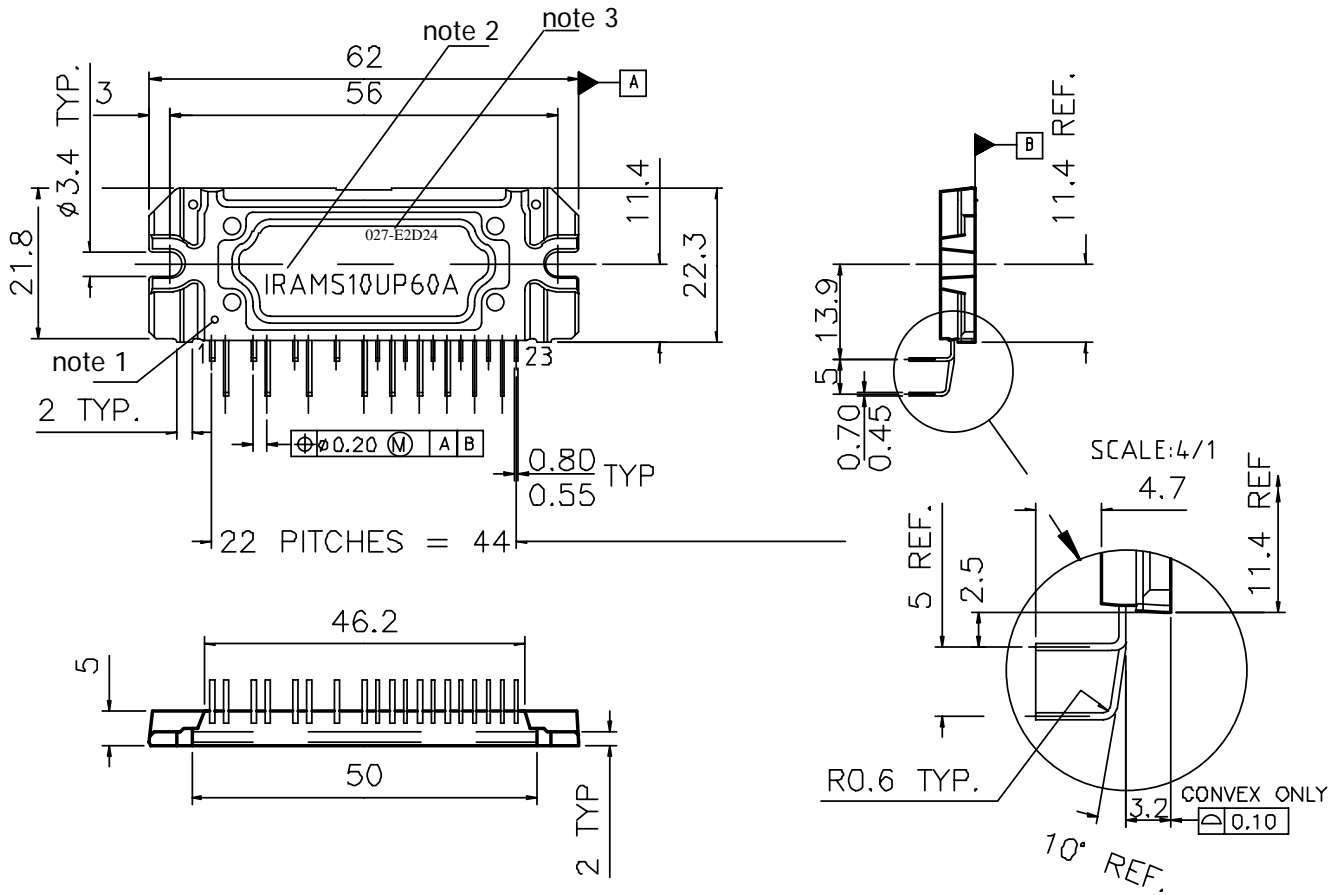
Dimensions in mm

- 1 - Marking for pin 1 identification
- 2- Product Part Number
- 3- Lot and Date code marking

For mounting instruction see AN-1049



## Package Outline



### Pin leadforming option -2

Notes:

Dimensions in mm

- 1 - Marking for pin 1 identification
- 2- Product Part Number
- 3- Lot and Date code marking

Data and Specifications are subject to change without notice